

PCI Express* 3.0 Technology: PHY Implementation Considerations for Intel Platforms

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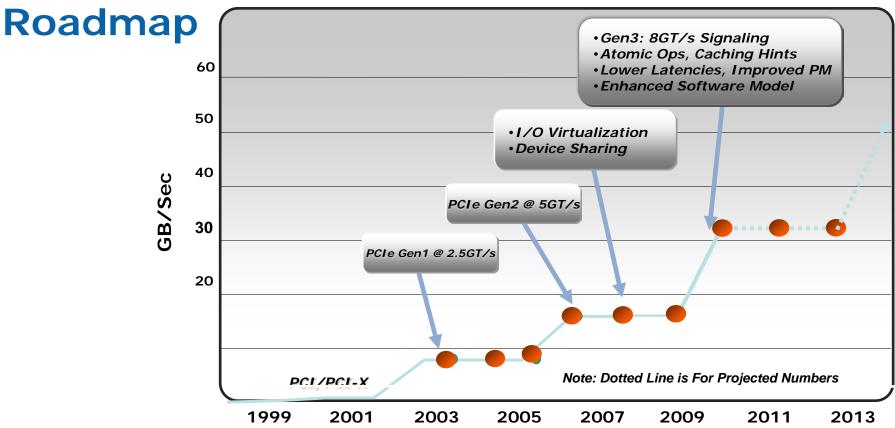
TCIS007



- Problem Statement
- Existing Usage of K-Code in 8b/10b
- Encoding Scheme
- Transmitter Equalization and training
- Implementation Considerations
- Summary



PCI Express* (PCIe*) Technology



	Raw Bit Rate	Link BW	BW/lane /way	BW x16
PCIe 1.x	2.5GT/s	2Gb/s	~250MB/s	~8GB/s
PCIe 2.0	5.0GT/s	4Gb/s	~500MB/s	~16GB/s
PCIe 3.0	8.0GT/s	8Gb/s	~1GB/s	~32GB/s

Based on x16 PCIe channel



Problem Statement

- PCI Express* (PCIe*) 3.0 data rate decision: 8 GT/s
 - High Volume Manufacturing channel for client/ servers
 - Same channels and length for backwards compatibility
 - Low power and ease of design
 - Avoid using complicated receiver equalization, etc.
- Requirement: Double Bandwidth from Gen 2
 - PCIe 1.0a data rate: 2.5 GT/s
 - PCIe 2.0 data rate: 5 GT/s
 - Doubled the data rate/ bandwidth from Gen 1 to Gen 2
 - Data rate gives us a 60% boost in bandwidth
 - Rest will come from Encoding
 - Replace 8b/10b encoding with a scrambling-only encoding scheme when operating at PCIe 3.0 data rate
- Double B/W: Encoding efficiency 1.25 X data rate 1.6 = 2X

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Review of K-Code Usage

- Each K-codes is a unique 10-bit value
 - Distinct from data and other K-codes
- Two flavors for K-code use
 - Packet Stream (independent of link width)
 - Lane Stream (per-lane)
- Packet Stream relates to Packet Framing (Link-Wide)
 - STP Start of Transaction Layer Packet (TLP)
 - END End (Good) of TLP
 - EDB End Bad of TLP
 - SDP Start of Data Link Layer packet (DLLP)
- Lane Stream relates to Ordered Sets:
 - Training Set #1 & #2: Training/ retraining
 - SKP Ordered Sets: clock compensation and byte realignment
 - Electrical Idle Start/ Exit sequence: Power Management
- New encoding scheme accommodates these existing usages

Functionality of K-Code needs to be preserved

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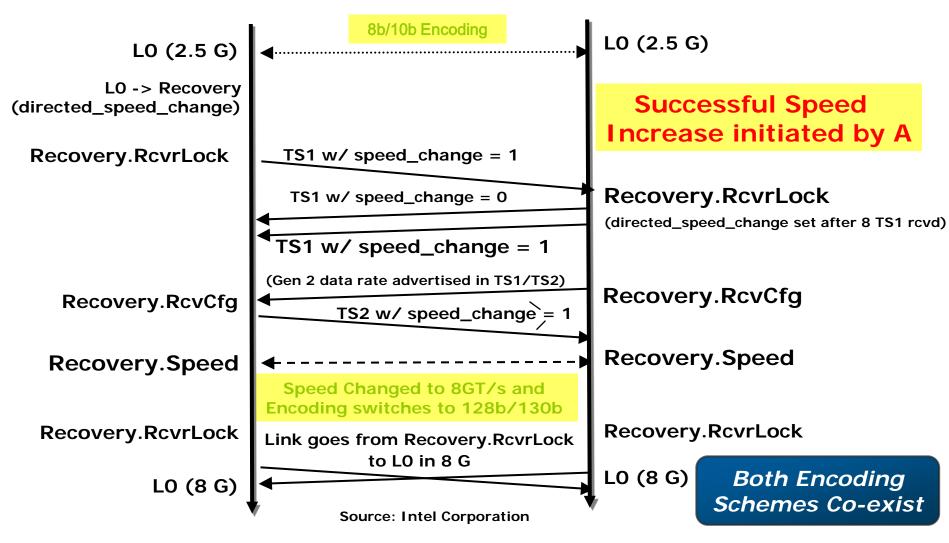
Requirements and Capabilities

- Basic Fault Model:
 - Guaranteed error detection against random bit flips in any packet or Ordered Set
- Eventual recovery from bit slip/add
- Handle killer packets
 - Send a different bit stream on retry of a packet
- Low bandwidth overhead (1-2%)
- Low L0s/L1 exit latency overhead
 - Preserve aggressive power management with performance
- Changes mostly limited to physical layer
- Protocol development concurrent with analysis / simulation done by Intel Pathfinding team

New encoding scheme: Better Performance and Reliability than PCI Express* 2.0 Technology

LTSSM Speed Change: Example

LTSSM in Device A LTSSM in Device B

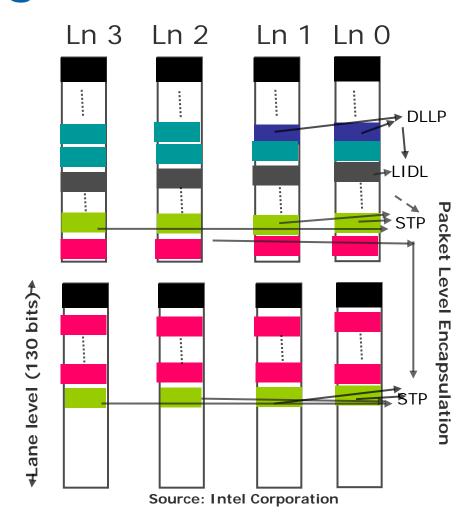




128b/130b Encoding Scheme

Two levels of encapsulation

- Lane Level: Blocks
 - Data vs Ordered Sets
 - 2-bit Sync Header identifies Data Block vs OS (not scrambled)
 - 128-bit payload
 - Rationale:
 - Redundancy helps separate Data from OS
 - 128-bit payload chosen to match OS payload. 2-bit is low overhead for Sync header is low
- Data Block: Link wide with Framing preamble identifying packet boundary up-front
 - Multiple packets within a Data Block and one packet can straddle multiple Blocks
 - Framing preamble same overhead as in 8b/10b
 - Payload scrambled

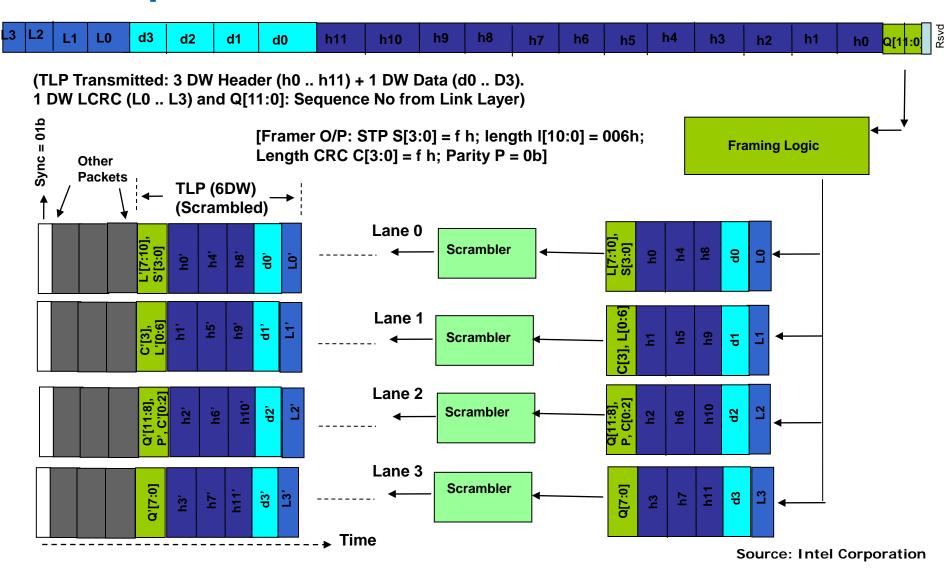




Data Block

- Packets: Logical IDL (LIDL), DLLP, TLP, etc.
- Various sizes from 8b/10b time: 1 Symbol for LIDL, DLLP: 8 Symbols, TLP: Multiple of 4 Symbols
- Everything other than TLP is fixed size
- Need to ensure triple bit flip detection ability while keeping the sizes the same
 - New encoding: One Symbol becomes 8bit rather than 10bit in 8b/10b
- TLP and DLLP body is LCRC protected which provides triple bit flip detection ability
 - Framing preamble itself needs to ensure triple bit flip detection ability as it is used to determine packet boundary
- Challenge: Framing preamble itself has to be of variable length
- Solution: Use first Symbol encoding between different entities to be at a Hamming distance of 4 to ensure triple bit flip detection ability
 - Subsequent Symbols, if any, use some form of CRC protection itself
- Robustness features confirmed by analysis/ simulations in Intel path finding

Example of TLP Transmission in a X4





Ordered Sets

- Used for Link training, power management, clock compensation
- Additional usage: Block alignment
 - Can not scrambled
 - Must reset scrambler (so that both sides start at same point)
 - Can not do this during Data Blocks
 - Unlike 8b/10b where COM can never be aliased to between any two 10bit Symbols, one can easily alias to any bit pattern when two Data Blocks are looked at consecutively for a fixed pattern
 - Must be done when all permutations are not possible (i.e., during Recovery)
 - Choice of Ordered Set encoding to be such that one can always correctly do block alignment
 - Another challenge is bit slip initially, based on past observation
 - Solution: Continuously do block alignment in Recovery while Ordered Sets are on
- Ordered Set for Clock compensation
 - Can not be scrambled (e.g., repeaters)
 - Need to carry information such as LFSR value to help trace tools
- Link training needs a spectrally rich pattern for better bit lock
- Solution: Most of TS1/TS2 are scrambled; Rest are not
- Problem scenarios and their solutions created and verified in analysis / simulation in Intel pathfinding

Sync Header Protection

- Challenge: Sync Header is 2-bits: so even a two bit flip can
 potentially change a Data Block to an Ordered Set and viceversa. Even worse the LFSR can be out of sync between the
 Tx and the Rx and introduce massive errors as a result and
 cause data corruption
- Solution: Define a "marker" of the appropriate type whenever there is a transition from Data Block to Ordered Set and vice-versa
 - Pre-notification of the Block type change that itself is adequately protected
 - Data Block to OS: In last Data Block as a CRC-protected packet
 - OS to Data Block: As an additional 130-bit marker OS
 - Protects more than triple bit flip
 - Problem scenario as well as solution created and validated through analysis/ simulation (Intel path finding)



Error Recovery

Framing error detected by PHY

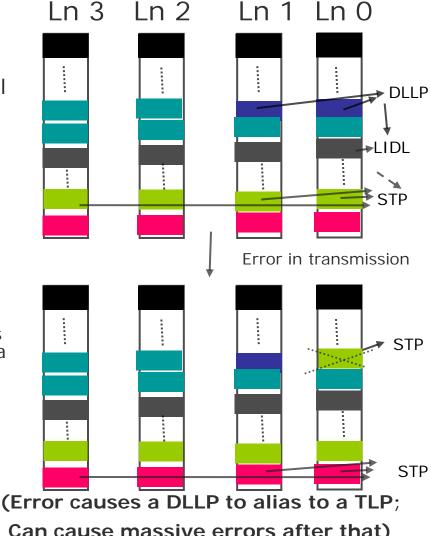
 Helps identify cases where the physical layer can either have its scrambler out of sync or fails to ascertain the next packet's framing preamble location

Any framing error directs LTSSM to Recovery

- Stop processing any received TLP/ DLLP after Recovery to avoid data corruption
 - The CRC within these packets become ineffective when the packet boundary is lost – random data can always alias to a good CRC
- Block lock and scrambler reset happens through Recovery prior to packet being accepted
- Link layer detected errors can be recovered through packet retry

Error Detection Guarantees maintained

Triple bit flip detection within each TLP/ DLLP/ IDL/ OS



Can cause massive errors after that)
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Transmitter Equalization

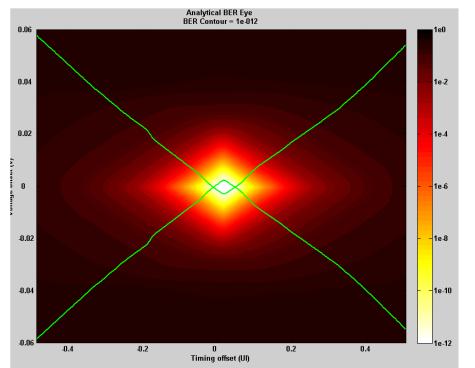
- 2.5 GT/s: Same de-emphasis for all
- 5 GT/s: Introduced platform dependent de-emphasis selection on a per-Link basis
 - -3.5 dB and -6 dB
- 8 GT/s: Our analysis shows that a static selection does not work for all channels due to variations in the receiver design, channel, as well as PVT
- Solution: Need to adjust each transmitter at the by its corresponding receiver in a fine-grain fashion (coefficients)
 - Need to do it once and store it for use on every entry to 8GT/s
 - Must start with some predefined value set by platform characteristics
 - Dynamic adjustment after that
 - Our analysis shows this approach results in working silicon



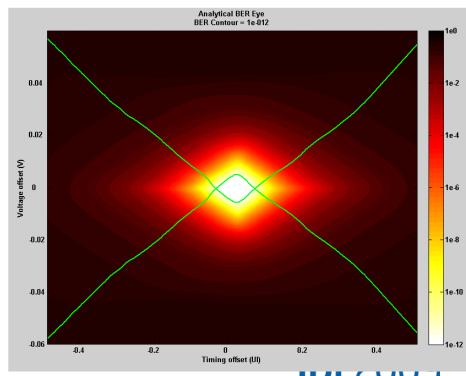
Tx EQ coefficient Optimization vs. Pre-set example

- The eye diagram on the left was the result of using the best pre-set Tx EQ values.
- The eye diagram on the right was same channel with optimized Tx EQ coefficients.
- The green contour shows the BER eye at 1e-12.
- Eye width opening increased from 7ps to 16ps (over 50% more Eye Width)
 - Both assumed a Tx EQ step size resolution of 1/32
 - Channel: 2 connector topology 18" pin-pin
 - Both used same Rx EQ that was re-optimized for each case.

BER Eye With Best Pre-Set



BER Eye with optimized Tx coef



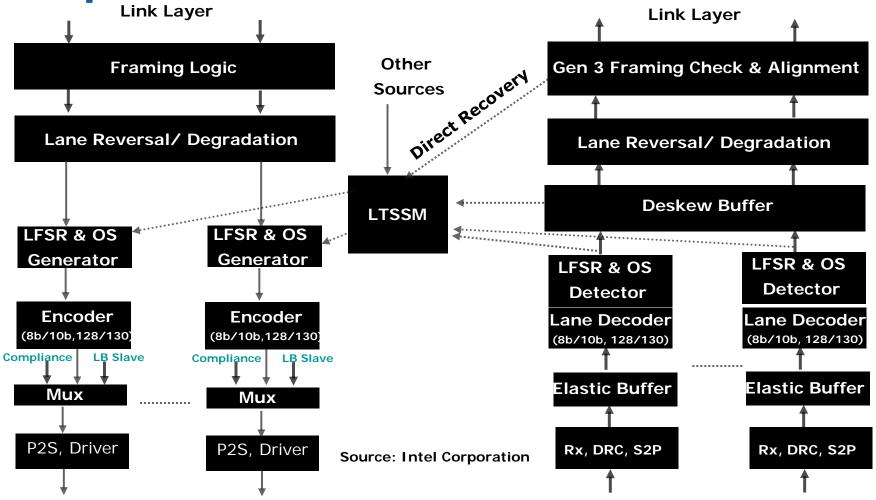
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Sample Transmitter and Receiver



Considerations: 1 byte offset to Link Layer for TLPs (EDB). Seq # aligned to LCRC



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Summary & Call to Action

- Overview of Logical PHY based on Intel analysis, simulations, and experience
- 128b/130b Encoding definition
- Equalization mechanism needed
- 25% bandwidth advantage with new encoding over 8b/10b encoding with enhanced reliability
- Track the PCI Express* (PCIe*) 3.0 Spec development in the PCI-SIG and at www.pcisig.com
- Track the PCIe PIPE Spec development at <u>www.intel.com/technology/pciexpress/devnet</u>
 - Plan for products accordingly



Additional Sources of Information on This Topic

- Other Sessions / Chalk Talks / Labs:
 - TCIQ002 Q&A: PCI Express* 3.0 Technology
 - TCISOO6 PCI Express* 3.0 Technology: Device Architecture optimizations on Intel Platforms
 - TCISOO8 Electrical requirements for designing PCIe* 3.0 ASICs on Intel platforms
 - USBS002 USB 3.0 Architecture and PHY Interface (PIPE)
 Specification Updates
- Demo/Booths:
 - PCI Express* Technology Community
- Additional Web-based Info:
 - www.pcisig.com
 - www.intel.com/technology/pciexpress/devnet



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